Speed-Boosted Adaptation and Applications to Swarms and Clusters of High-Performance Aerospace Systems

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Control of Agile Responsive Aerospace Systems





Aleksandyr Mikhailovich Lyapunov

Lyapunov-like techniques are usually the basis of most nonlinear stability analysis

- Controllers are synthesized to suit "chosen" Lyapunov functions
- Lyapunov's Direct Method paired with LaSalle invariance, Barbalat's lemma establish foundations for optimal, robust, and adaptive control
- Finding the right Lyapunov function is more art than science



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$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= W(x)\theta^* + u \\ \end{aligned} \qquad \left\{ \begin{array}{c} u = -k_p x_1 - k_v x_2 - W(x)\theta^* \\ k_p > 0, \ k_v > 0 \\ \end{array} \right.$$

The closed-loop system is UES. This can be established via

$$A_m \doteq \begin{bmatrix} 0 & 1 \\ -k_p & -k_v \end{bmatrix}, \quad A_m^T P + P A_m = -Q, \quad V_1 = \mathbf{x}^T P \mathbf{x} \to \dot{V}_1 = -\mathbf{x}^T Q \mathbf{x} < 0$$

Alternately, we could consider "energy-like" function

$$V_2 = (k_p x_1^2 + x_2^2)/2 \quad \rightarrow \quad \dot{V}_2 = -k_v x_2^2 \le 0$$

Thus, $V_2(x)$ is "non-strict" (aka *defective*) but the story still has a happy ending, thanks to LaSalle Invariance, Barbalat's Lemma...

Questions: How do we construct strict Lyapunov functions? Why bother about them?

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Constructing Strict Lyapunov Functions

- If UGAS is already known, converse theory guarantees existence
- Explicit availability of a strict Lyapunov function aids robustness analysis (external disturbances, adaptive control, time-delays, ...)
- Construction is a challenging problem, significant ongoing research (Mazenc, Malisoff, Teel, Nesic, etc.)



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- Higher-order Lie derivatives of non-strict Lyapunov functions
- Use of continuous-time Matrosov theorem
- Feedback with small gains
- Sufficient conditions, usually non-quadratic functions

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Strictification via State-Dependent Switching





Angular Velocity Observer Application:

$$\begin{aligned} & \text{Relative Orientation} \\ & e_q = \left[\begin{array}{c} \dot{q}_0 q_v + q_0 \hat{q}_v + q_v^{\times} \hat{q}_v \\ & q_0 \hat{q}_0 - q_v^{\text{T}} \hat{q}_v \end{array} \right] \\ & \\ & \mathbf{e}_{\omega} = \omega - \mathbf{C}(\mathbf{e}_{\mathbf{q}}) \hat{\omega} - \mathbf{e}_{\mathbf{q}_v} \end{aligned}$$

- Salcudian 1991, Open Problem (till Chunodkar, Akella, JGCD 2014)
- Switching provides strictification while ensuring C⁰ continuity of states
- Finite number of switches no zeno-type behavior
- Smooth analog for this result available through a spiral design approach (Thakur, Mazenc, Akella, JGCD 2015)



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Selecting the filter gain $\alpha = (k_p + k_v)$ results in $\dot{V}_3 = -k_p x_1^2 - k_v x_{2f}^2$

Mr. Lyapunov is both QUADRATIC and STRICT again!

The control signal can be recovered by

$$u = \dot{u}_f + \alpha u_f \implies u = -\alpha k_p x_1 - (k_p + k_v) x_2 - W(\mathbf{x}) \theta^*$$

Thus, for this academic example, filters are for analysis ONLY and they aren't needed for implementation!

Introduce stable linear low-pass filters (i.e., $\alpha > 0$)

$$\dot{x}_{1f} = -\alpha x_{1f} + x_1, \qquad \dot{x}_{2f} = -\alpha x_{2f} + x_2 \dot{u}_f = -\alpha u_f + u, \qquad \dot{W}_f = -\alpha W_f + W(\mathbf{x})$$

Simple algebra results in the following, modulo exponentially decaying terms,

$$\begin{aligned} \dot{x}_{1f} &= x_{2f} \\ \dot{x}_{2f} &= W_f \theta^* + u_f \end{aligned} \begin{cases} u_f = -k_p x_1 - k_v x_{2f} - W_f \theta^* \\ k_p > 0, \ k_v > 0 \end{aligned}$$

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Big Trouble! We are staring at the Uniform Detectability Obstacle

Fix: Either introduce non-intuitive cross-terms to "strictify" the Lyapunov function or, possibly adopt the filter embedment approach

And.. this is only the tip of the iceberg..

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- Mature subject area
- Several variants exist
 - Direct/Indirect
 - Backstepping
 - Immersion & Invariance
 - *L*₁ Adaptive Control



Procustes' Mythical Bed

- ► Fact 1: Even a linear plant under the action of an adaptive controller becomes nonlinear in the closed-loop due to the adaptation mechanism
- Fact 2: Plant parameters affine in the governing dynamic model
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Consider a prototypical adaptive stabilization problem

Suppose all plant parameters θ^* are known and

 $\boldsymbol{u} = \boldsymbol{k}(t)\boldsymbol{h}(\boldsymbol{x}) + \boldsymbol{W}(\boldsymbol{x})\boldsymbol{\theta}^*$

is the controller that achieves the desired control objective

• Then, in the case θ^* is unknown, design controller

 $\mathbf{u} = \mathbf{k}(t)\mathbf{h}(\mathbf{x}) + W(\mathbf{x})\hat{\mathbf{\theta}}(t)$

together with a suitable update law of $\hat{\theta}(t)$ (parameter estimator) so that the closed-loop is stable and the control objective is again achieved.

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• Add tuning function $\beta(x)$ to the adaptation (Astolfi & Ortega, IEEE TAC, 2003) :

 $\boldsymbol{u} = \mathbf{k}(t)h(\boldsymbol{x}) + W(\boldsymbol{x})\left[\hat{\theta}(t) + \boldsymbol{\beta}(\boldsymbol{x})\right]$

The state-dependent tuning function β(x) should satisfy an integrability condition:

$$\left[\frac{\partial \beta(x)}{\partial x}\right]^{T} W(x) + W^{T}(x) \frac{\partial \beta(x)}{\partial x} = Q(x) \ge 0 \text{ uniformly in } x$$

- Sufficient condition ONLY (... think $A^T P + PA = -Q$)
- Q(x) is a design function
- β is not uniquely defined
- Affine uncertainty representation not necessary
- Nonlinear single-input systems in cascade form and linear multi-input systems always satisfy the manifold attractivity condition (Akella, Subbarao, SCL 2005)
- Stability analysis:

$$V = \frac{1}{2}\mathbf{x}^{\mathsf{T}}\mathbf{x} + \frac{\sigma}{2}\mathbf{z}^{\mathsf{T}}\mathbf{z}, \qquad \mathbf{z} \doteq \hat{\theta} - \theta^* + \beta, \quad \sigma > 0,$$

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$$\dot{\mathbf{x}} = -\mathbf{x} - \mathbf{W}(\mathbf{x})\tilde{\mathbf{\theta}}; \quad \{\tilde{\mathbf{\theta}} = \hat{\mathbf{\theta}} - \mathbf{\theta}^* \\ \dot{\hat{\mathbf{\theta}}} = \gamma_{ce} \mathbf{W}^{\mathsf{T}}(\mathbf{x})\mathbf{x}$$

- Performance ultimately dictated by parameter estimator $\Rightarrow W(\mathbf{x})\tilde{\theta}$ like disturbance
- Parameter estimates driven by the regulating/tracking error
- Unable to mimic $\dot{x} = -x$

The attracting manifold design, on the other hand, results in

$$\begin{aligned} \dot{x} &= -x - W(x)z \\ \dot{z} &= -\gamma W^{\mathsf{T}}(x)W(x)z \end{aligned} \begin{cases} \lim_{t \to \infty} W(x)z = 0; \quad z = \hat{\theta} + \beta - \theta^* \\ \dot{z} = 0 \text{ for all } t > t^* \text{ if } z(t^*) = 0 \end{aligned}$$

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- Permits non-strict Lyapunov Functions (bypass detectability obstacle)
- Of course requires satisfaction of the integrability condition

CE based designs typically result in

$$\begin{aligned} \dot{\mathbf{x}} &= -\mathbf{x} - \mathbf{W}(\mathbf{x}) \tilde{\mathbf{\theta}}; \quad \{ \tilde{\mathbf{\theta}} = \hat{\mathbf{\theta}} - \mathbf{\theta}^* \\ \dot{\hat{\mathbf{\theta}}} &= \gamma_{ce} \mathbf{W}^{\mathsf{T}}(\mathbf{x}) \mathbf{x} \end{aligned}$$

- Performance ultimately dictated by parameter estimator $\Rightarrow W(\mathbf{x})\tilde{\theta}$ like disturbance
- Parameter estimates driven by the regulating/tracking error
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Speed-Boosted Adaptation

Wash all states and the regressor through stable linear low-pass filters

- The regressor filter assures circumvention of the integrability obstacle. Specifically, $\beta = W_f^T x_f$ satisfies the integrability condition
- The closed-loop system becomes

$$\dot{\mathbf{x}}_{f} = -\mathbf{x}_{f} - W_{f}(t)\mathbf{z} \dot{\mathbf{z}} = -\gamma W_{f}^{T} W_{f} \mathbf{z}$$

- ▶ Very high-dimensional closed-loop system ($x_f \in \mathcal{R}^n$; $W_f \in \mathcal{R}^{n \times p}$)
- ► Speed boosting: $\mathbf{k}(t) = k\rho(t)r^2(t)$ k > 0, $\inf_{t \ge 0} \rho(t) = \rho^* > 0$
 - ▶ Scalar extension: (non-filter, ho(t) = 1); $\mathbf{k} \in \mathcal{L}_2 \cap \mathcal{L}_\infty$ (Yang, Akella, SES 2015)
 - Second-order dynamic extension (Filter)
 - $\dot{
 ho}=-(
 ho-1/r^2-arepsilon), \quad
 ho(0)=1/r^2(0)+arepsilon, 0<arepsilon\ll 1$
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Generalizations shown to hold for Euler-Lagrange class of systems (Yang, Akella, Mazenc, ACC 2016)

 $M(q)\ddot{q} + C(q, \dot{q})\dot{q} + F\dot{q} + G(q) = u$

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Adaptive Attitude Tracking Application

System parameters and reference rate profile

(Seo, Akella, JGCD 2008)

$$J = \left[\begin{array}{rrrr} 20 & 1.2 & 0.9 \\ 1.2 & 17 & 1.4 \\ 0.9 & 1.4 & 15 \end{array} \right]$$

 $\omega_r(t) = 0.3(1 - e^{-0.01t^2})\cos t + te^{-0.01t^2}(0.08\pi + 0.006\sin t)$

• Tracking errors and dynamic gain $\mathbf{k}(t) = \eta_1(t)$



(Yang, Akella, AIAA/AAS SFM, 2016)

Simulation Results

Control and estimation error norms



• Large initial rate error: $\|\delta \omega(0)\| = \sqrt{3}$



Coordinated Sensing and Decentralized Control

Distributed Heterogeneous Networks:

- Mission ' tas decomposition
 Minimal communication, persistenct
- Coordination, optimality and constraint satisfaction



Research Focus:

- Self organization clustering
- GPS-denied navigation, path-planning

Consensus establishment Time-delay in communication

- Undirected/Symmetric
- Rigid, but not minimally so



- Directed/Asymmetric
- Minimally persistent
- LFF, LRF, Co-Leader



Clustering for Self Organization

Hierarchial Self-Organization of the Network

- Determination of clusterheads and clients
- Optimization a very difficult problem (NP hard)
- Best approximations ~ $O(\log n)$ for 1-D; $O(\sqrt{n})$ for 2-D
- Mobile nodes not involved





August 17, 2016

Time-Delays & Imperfect Communication



Dynamics with Unstable Drift:

- Graph containing spanning tree necessary for consensus
- Necessary and sufficient stability conditions for *cyclic graphs* in terms of control gains α (position-feedback) and β (rate-feedback)
- Directed graphs are *less robust* w.r.t. time-delay uncertainty when compared to corresponding underlying undirected graphs

Time-Delays & Imperfect Communication



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Stability Conditions - No Self-Delay Protocol

No self-delay, weighted adjacency matrix \bar{A}

• Critical delay $\tau^* \leq \tau_{\max}$

(Yang, Mazenc, Akella, JGCD 2015)



Stability Conditions - with Same Self-Delay

- Same self-delay, weighted adjacency matrix \bar{A}
- Critical delay $\tau^* \leq \tau_{\max}$



Swarm about Dwarf Planet Ceres (Hernandez, Thakur, Akella, JGCD 2015)





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Navigation without Localization/GPS Infrastructure





GPS Denied Robot Navigation:

- Reach "purple" from "blue"
- Arbitrary heading
- Imperfect communication boundaries





Vision-Based Discrete Adaptive Rate Estimation (Almeida, Akella, Mortari, AIAA/AAS SFM 2016)



What if Landmarks aren't Mapped?

Simultaneous Localization and Mapping (15 fps) (OrbSLAM; Mur-Artal et al. Universidad Zaragoza 2016)



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Challenges, Opportunities, Future Work..

- Distributed, ubiquitous
- "Internet of Things" at massive scales
- Human-robot interface, perception, cognition





- Layered-autonomy, dependability
- Uncertainty, quantification and its impact on sensor/resource allocation



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Thank you. Questions?

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